

DSP First, 2/e

Lecture 18 DFS: Discrete Fourier Series, and Windowing

READING ASSIGNMENTS

- This Lecture:
 - Chapter 8, Sections 8-3, 8-5 & 8-6

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LECTURE OBJECTIVES

- **Discrete Fourier Series** for periodic $x[n]$

$$a_k = \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-j(2\pi/N)kn}$$

$$x[n] = \sum_{k=0}^{N-1} a_k e^{j(2\pi/N)kn}$$

- DFT of one period with scaling by $1/N$ gives scaled DFS coefficients
- **Windowing**
 - extract short sections from long signal

Review

- **Discrete Fourier Transform (DFT)**

$$X[k] = \sum_{n=0}^{N-1} x[n] e^{-j(2\pi/N)kn} \quad x[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{j(2\pi/N)kn}$$

- DFT is **frequency sampled** DTFT
 - For finite-length signals
- DFT computation via FFT
 - FFT of zero-padded signal \rightarrow more freq samples
- Transform pairs & properties (DTFT & DFT)

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Comparison: DFT and DTFT

Discrete Fourier Transform (DFT)

$$X[k] = \sum_{n=0}^{N-1} x[n] e^{-j(2\pi/N)kn}$$

Inverse DFT

$$x[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{j(2\pi/N)kn}$$

Discrete-time Fourier Transform (DTFT)

$$X(e^{j\hat{\omega}}) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\hat{\omega}n}$$

Inverse DTFT

$$x[n] = \frac{1}{2\pi} \int_0^{2\pi} X(e^{j\hat{\omega}}) e^{j\hat{\omega}n} d\hat{\omega}$$

Inverse DFT always makes a periodic signal

- Evaluate N-pt IDFT outside of [0,N-1]

$$\begin{aligned} x[n+N] &= \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{j(2\pi/N)k(n+N)} \\ &= \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{j(2\pi/N)kn} e^{j(2\pi/N)k(N)} \\ &= \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{j(2\pi/N)kn} = x[n] \end{aligned}$$

- Thus the IDFT synthesizes a periodic signal

Fourier Series for Discrete-Time Signal

Given a periodic sequence $x[n]$, how do we write it as a sum of sinusoids (or complex exponentials) ?

Which frequencies? How many? Fundamental ?

Exponentials must have the same period as $x[n]$, which is N . There are only N possible exps.

$$\left\{ e^{j(2\pi k/N)n} \right\}_{k=0}^{N-1} \text{ for } n = 0, 1, \dots, N-1$$

\downarrow \downarrow
 $e^{j(0)n}, e^{j(2\pi/N)n}, e^{j(4\pi/N)n}, e^{j(6\pi/N)n}, \dots, e^{j(2\pi(N)/N)n}$

Discrete Fourier Series Representation (2)

$$\left\{ e^{j(2\pi k/N)n} \right\}_{k=0}^{N-1} \text{ for } n = 0, 1, \dots, N-1$$

$$x[n] = \sum_{k=0}^{N-1} a_k e^{j(2\pi k/N)n}$$

Given the sequence $x[n]$, how do we find a_k ?

Recall IDFT always synthesizes a periodic $x[n]$

$$\begin{aligned} x[n] &= \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{j(2\pi k/N)n} \\ \Rightarrow a_k &= \frac{1}{N} X[k] \end{aligned}$$

So, we find a_k by taking the N -pt DFT of one period of $x[n]$ and then multiplying by $1/N$

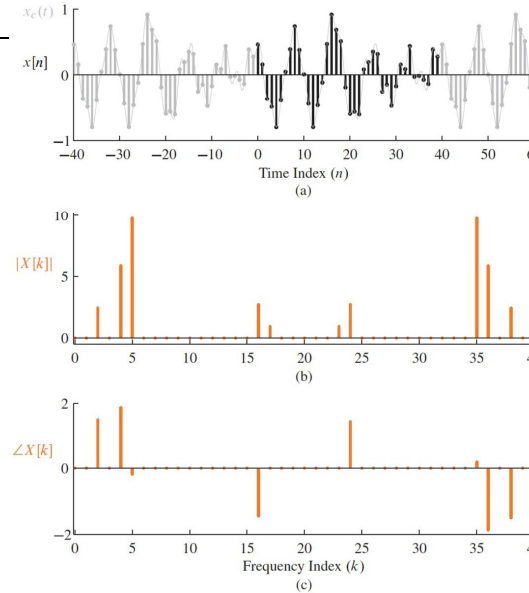
Discrete Fourier Series (DFS)

$$x[n] = \sum_{k=0}^{N-1} a_k e^{j(2\pi k/N)n} \leftarrow \text{Synthesis of a periodic signal } x[n] = x[n+N]$$

$$a_k = \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-j(2\pi k/N)n}$$

Find a_k by taking N-pt DFT of one period of $x[n]$ and then multiplying by $1/N$

DFT of one period → DFS



Can you get the DFS coeffs from this DFT ?

Figure 8-12 Periodic sequence $x[n]$ in (8.43) and the corresponding 40-point DFT spectrum $X[k]$. (a) Dark region indicates the 40-point interval (starting at $n = 0$) taken for analysis. (b) DFT magnitude spectrum. (c) DFT phase spectrum.

DFS Synthesis Example

$$x[n] = \sum_{k=0}^{N-1} a_k e^{j(2\pi k/N)n}$$

$$\{a_k\} = \{1, 1, 1, 0, 0, 0, 1, 1\} \text{ for } n = 0, 1, 2, 3, 4, 5, 6, 7$$

$$x[n] = e^{j(0)n} + e^{j(2\pi/8)n} + e^{j(4\pi/8)n} + e^{j(12\pi/8)n} + e^{j(14\pi/8)n}$$

Recall negative frequencies in high indices for the DFT

$$e^{j(14\pi/8)n} = e^{j((16\pi - 2\pi)/8)n} = e^{-j(2\pi/8)n}$$

$$\begin{aligned} x[n] &= e^{j(0)n} + e^{j(2\pi/8)n} + e^{j(4\pi/8)n} + e^{-j(4\pi/8)n} + e^{-j(2\pi/8)n} \\ &= 1 + 2\cos\left(\frac{\pi}{4}n\right) + 2\cos\left(\frac{\pi}{2}n\right) \end{aligned}$$

How is DFS related to cont-time Fourier Series ?

- Fourier Series (from Chap. 3)

$$x(t) = \sum_{k=-\infty}^{\infty} a_k e^{j(2\pi k f_0)t}, \text{ where } a_k = |a_k| e^{j\angle a_k}$$

- Need to obey the Nyquist rate: i.e., band-limited signal

$$x(t) = \sum_{k=-M}^M a_k e^{j(2\pi k f_0)t}, \text{ and } f_s > 2Mf_0$$

- Then sample

$$\begin{aligned} x[n] &= \sum_{k=-M}^M a_k e^{j(2\pi k f_0)(n/f_s)}, \text{ and } f_s > 2Mf_0 \\ &= \sum_{k=-M}^M a_k e^{j(2\pi k/N)n}, \text{ if } f_s = Nf_0 \end{aligned}$$

How is DFS related to cont-time Fourier Series (2)

- Band-limited Fourier Series (from Chap. 3)

$$x(t) = \sum_{k=-M}^M a_k e^{j(2\pi k f_0)t}, \text{ and } f_s > 2Mf_0$$

- Can be sampled to give periodic $x[n]$

$$x[n] = \sum_{k=-M}^M a_k e^{j(2\pi k/N)n}, \text{ if } f_s = Nf_0$$

- Compare to DFS

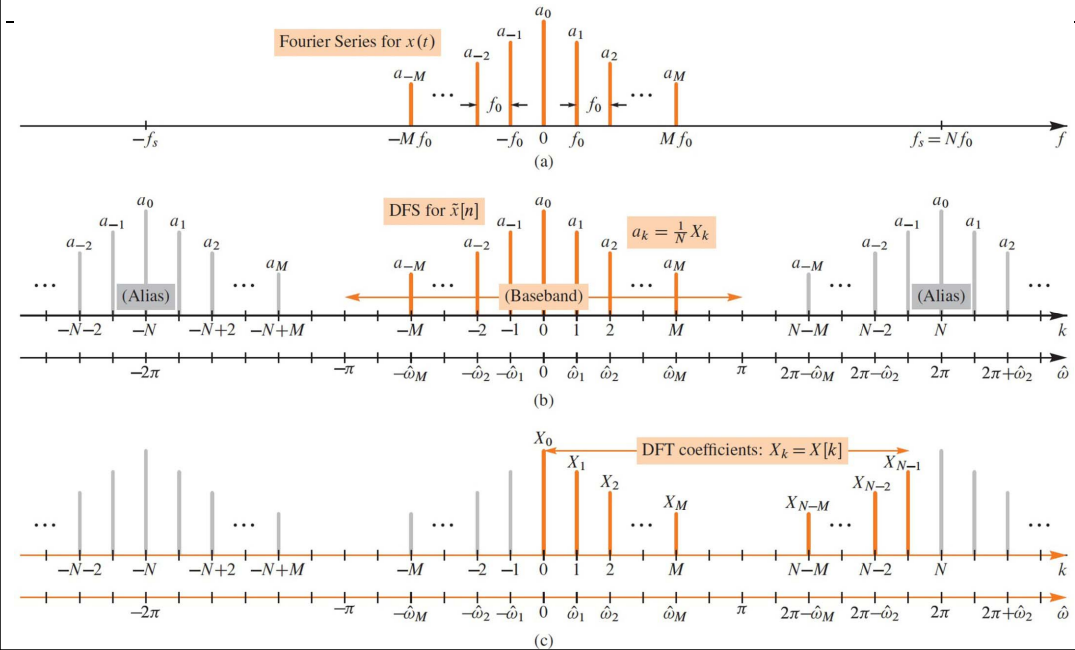
$$x[n] = \sum_{k=0}^{N-1} a_k e^{j(2\pi k/N)n}$$

- Same coefficients a_k

Recall

$$\hat{\omega} f_s = \omega$$

Spectrum Analysis of a Periodic Signal $x[n]$



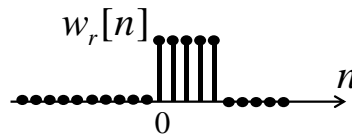
Windows

- Finite-Length signal (L) with positive values

- Extractor
- Truncator

Rectangular Window

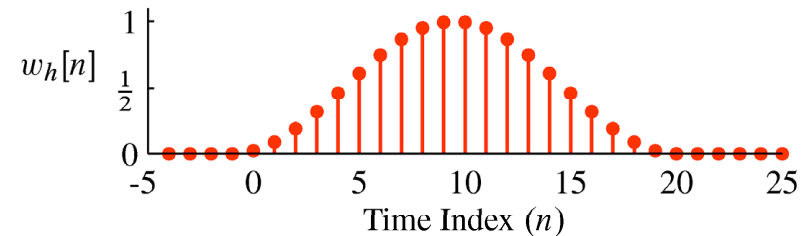
$$w_r[n] = \begin{cases} 0 & n < 0 \\ 1 & 0 \leq n < L \\ 0 & n \geq L \end{cases}$$



$$w_r[n]x[n+n_0] = \begin{cases} 0 & n < 0 \\ x[n+n_0] & 0 \leq n < L \\ 0 & n \geq L \end{cases}$$

Von Hann Window (Time Domain)

- Plot of Length-20 von Hann window

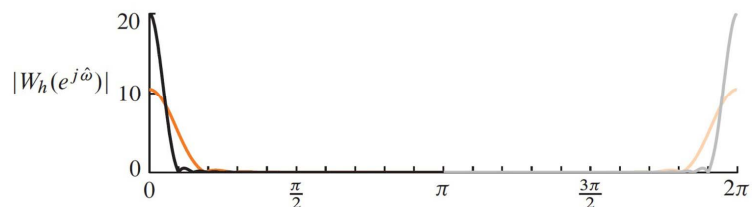


von Hann Window (Length L)

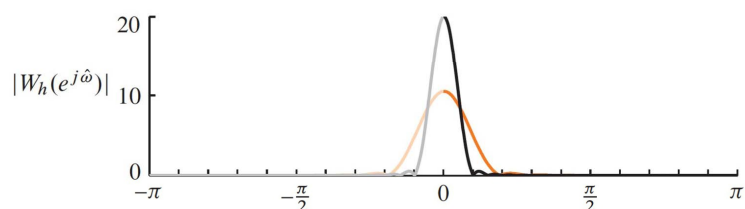
$$w_h[n] = \begin{cases} 0 & n < 0 \\ \frac{1}{2} - \frac{1}{2} \cos(2\pi(n+1)/(L+1)) & 0 \leq n < L \\ 0 & n \geq L \end{cases}$$

Von Hann Window (Frequency Domain)

- DTFT (magnitude) of Length-20 Hann window



(a)



Frequency (ω)

(b)

Window section of sinusoid, then DFT

- Multiply the very long sinusoid by a window

$$x[n] = A \cos(\hat{\omega}_0 n + \varphi) \quad -\infty < n < \infty$$

$$\hat{\omega}_0 = 0.4\pi \rightarrow$$

- Take the N-pt DFT
 - Finite number of frequencies (N)
 - Finite signal length (L) = window length

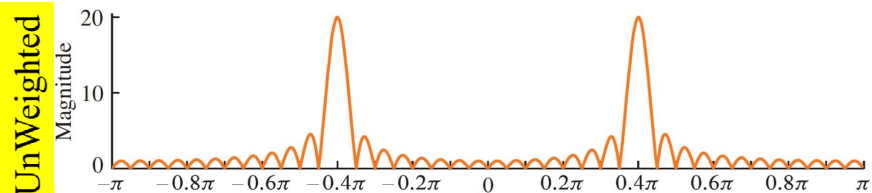
$$X_L(e^{j\hat{\omega}}) = \sum_{n=-\infty}^{\infty} w_L[n] x[n] e^{-j\hat{\omega}n} \rightarrow \sum_{n=0}^{L-1} w_L[n] x[n] e^{-j\hat{\omega}_k n}$$

$$\hat{\omega}_k = (2\pi / N)k, \quad k = 0, 1, 2, \dots, N-1$$

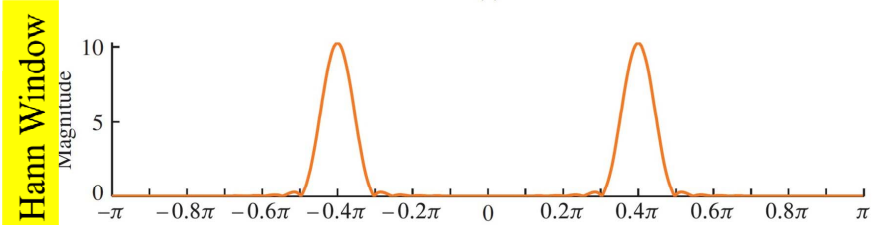
Expectation: 2 narrow spectrum lines →

DTFT of Windowed Sinusoid (with different windows)

- DTFT (magnitude) of windowed sinusoid
 - Length-40 Hann window vs Length-40 Rectangular window



(a)

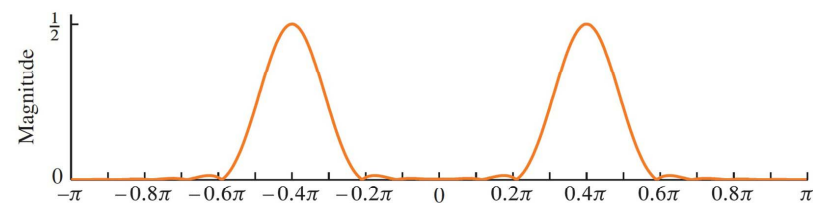


Frequency ($\hat{\omega}$)

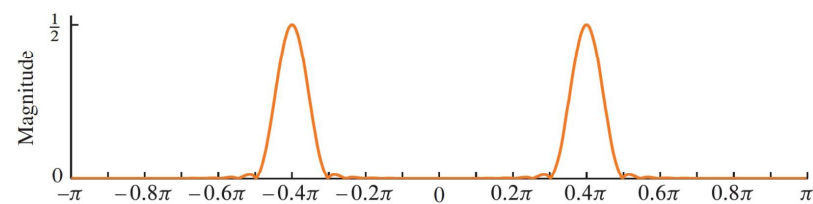
(b)

Change Window Length

- DTFT (magnitude) of windowed sinusoid.
 - Length-20 Hann window vs. Length-40 Hann window



(a)



Frequency ($\hat{\omega}$)

(b)