

DSP First, 2/e

Lecture 19
Spectrogram: Spectral
Analysis via DFT & DTFT

READING ASSIGNMENTS

- This Lecture:
 - Chapter 8, Sects. 8-6 & 8-7
- Other Reading:
 - FFT: Chapter 8, Sect. 8-8

LECTURE OBJECTIVES

- **Spectrogram**
- Time-Dependent Fourier Transform
 - DFTs of short windowed sections
- Frequency Resolution
 - To resolve closely spaced spectrum lines
 - Need long windows

Review & Recall

- **Discrete Fourier Transform (DFT)**

$$X[k] = \sum_{n=0}^{N-1} x[n] e^{-j(2\pi/N)kn} \quad x[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{j(2\pi/N)kn}$$

- DFT is **frequency sampled** DTFT
 - For finite-length signals
- DFT computation is actually done via FFT
 - FFT of zero-padded signal \rightarrow more freq samples
- Transform pairs & properties (DTFT & DFT)

Recall: DFT and DTFT

Discrete Fourier Transform (DFT)

$$X[k] = \sum_{n=0}^{N-1} x[n] e^{-j(2\pi/N)kn}$$

Inverse DFT

$$x[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{j(2\pi/N)kn}$$

Discrete-time Fourier Transform (DTFT)

$$X(e^{j\hat{\omega}}) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\hat{\omega}n}$$

Inverse DTFT

$$x[n] = \frac{1}{2\pi} \int_0^{2\pi} X(e^{j\hat{\omega}}) e^{j\hat{\omega}n} d\hat{\omega}$$

The Transform Method (DTFT)

- To get the output signal $y[n]$, given the input $x[n]$, and the system defined by $h[n]$.

$$y[n] = h[n] * x[n]$$

$$x[n] \rightarrow X(e^{j\hat{\omega}})$$

$$Y(e^{j\hat{\omega}}) = H(e^{j\hat{\omega}}) X(e^{j\hat{\omega}})$$

$$Y(e^{j\hat{\omega}}) \rightarrow y[n]$$



- Convolution becomes multiplication of DTFTs

Frequency Response Method

- To get the output signal $y[n]$, given the input $x[n]$ is a sinusoid, and the system defined by $h[n]$.

$$y[n] = h[n] * A \cos(\hat{\omega}_0 n + \varphi)$$

$$h[n] \rightarrow H(e^{j\hat{\omega}})$$

$$\text{eval } H(e^{j\hat{\omega}}) \rightarrow H(e^{j\hat{\omega}_0}) = H_0$$

$$y[n] = A \cdot |H_0| \cos(\hat{\omega}_0 n + \varphi + \angle H_0)$$

- Multiply Magnitudes, Add Phases

DTFT, DFT and DFS are Tools for Analysis

- DTFT** applies to discrete time-sequences, regardless of length (as long as $x[n]$ is absolutely summable);
 - Note: sinusoids are not absolutely summable
- N-pt DFT** of a finite-length sequence is equivalent to **sampling the corresponding DTFT** on frequency axis at N equally spaced points, **without losing any information**
 - Zero-padding** when $N > L$
- If the **sequence is periodic**, and the **DFT is performed on one period**, the result is a discrete Fourier Series (**DFS**) which is an exact representation.

Various Situations in Signal Analysis

A. Signal property does not fluctuate with time

1. Sampled signal is periodic with **known** period N_0 – use N_0 -pt DFT and get the precise result (DFS)
2. Sampled signal is aperiodic, or period is **unknown**, or hard to ascertain precisely.
3. Signal characteristic is global (for the entire signal)
 - a. Finite-Length, e.g., pulse
 - b. Infinite-Length: e.g., right-sided exponential

COMMON CASE →

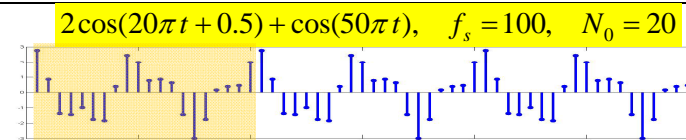
B. Sampled signal property changes with time

1. Signal defined by math formulas over intervals
2. **Recorded** signal, most likely with changing sinusoidal content, or properties – the most general case

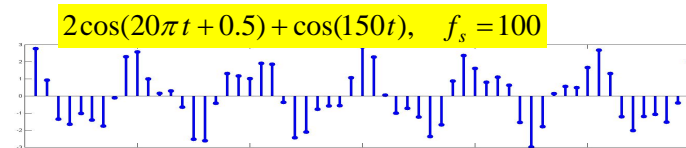
COMMON CASE →

Examples

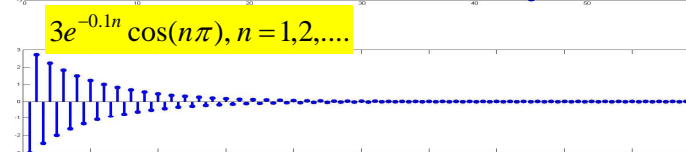
A.1 Periodic Sequence



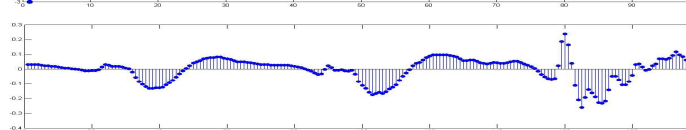
A.2 Nonperiodic Sequence



A.3 Global infinite-length sequence

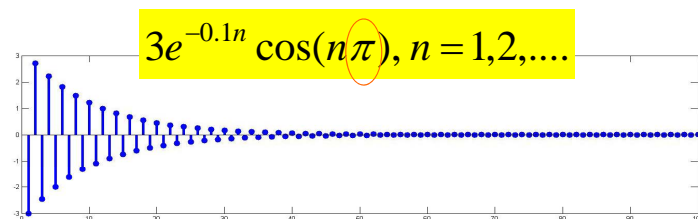


B.2 A general sequence

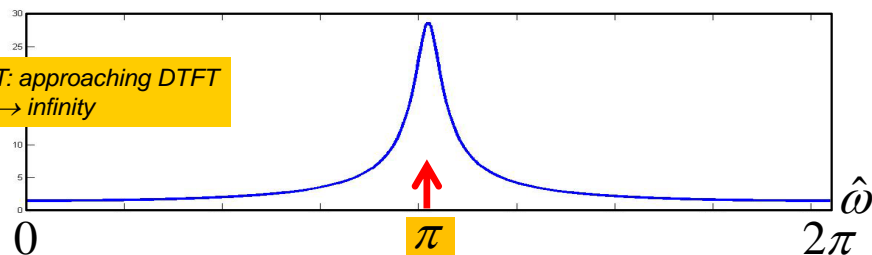


A.3: DTFT gives global spectrum

A.3 sequence from formula



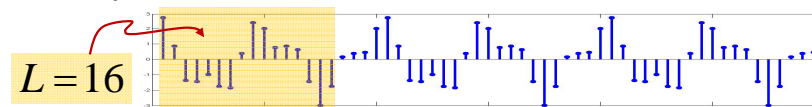
N -pt DFT: approaching DTFT when $N \rightarrow \infty$



More about this later.

When Periodicity Is Not Precise – A.2

- Periodicity is imprecise – either not known exactly or impractical to take exactly one period of data; for example:

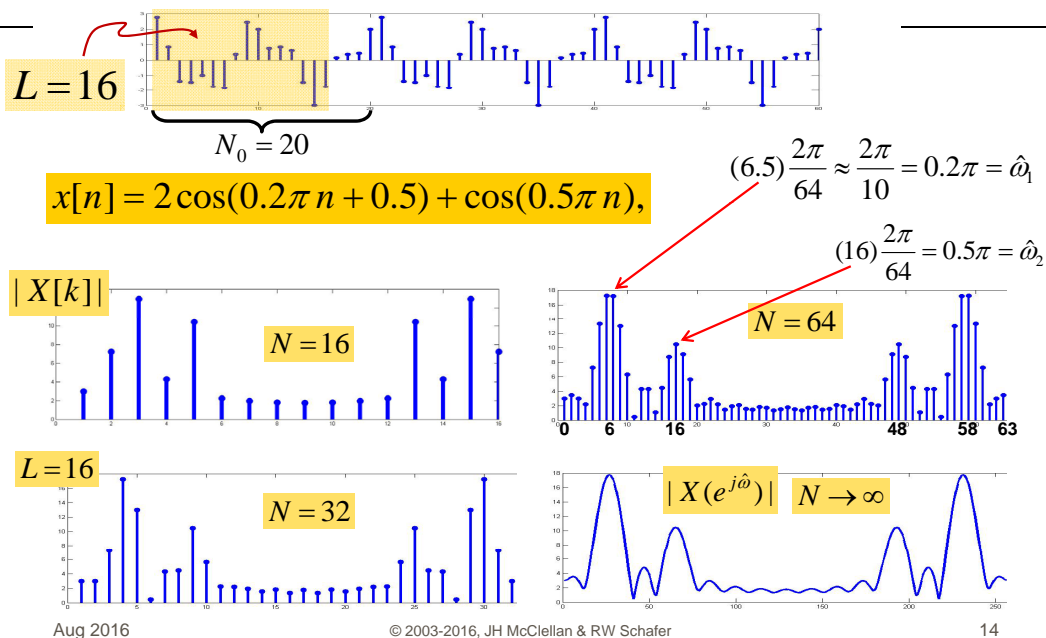


$N_0 = 20$ Actual period of sequence

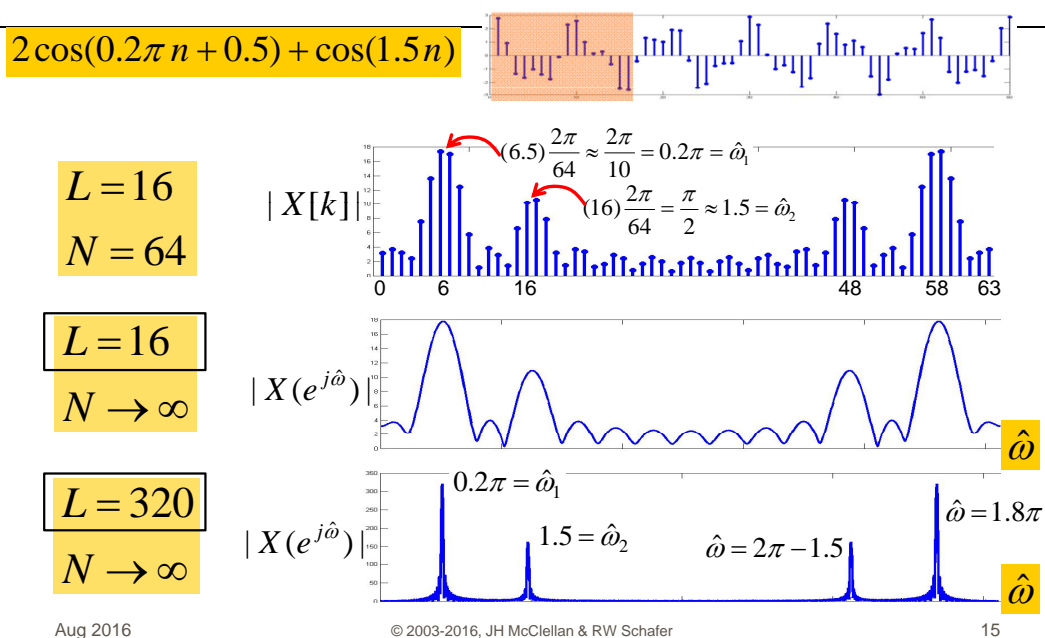
$x(t) = 2 \cos(20\pi t + 0.5) + \cos(50\pi t),$
 $f_0 = 5, f_s = 100, N_0 = 20$

- We take L data points for analysis; actual period is not L ($\neq 20$ for the example).

Imprecise Period Example (2)



Similar Example - Changing L



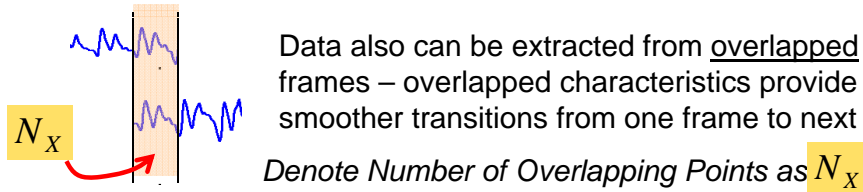
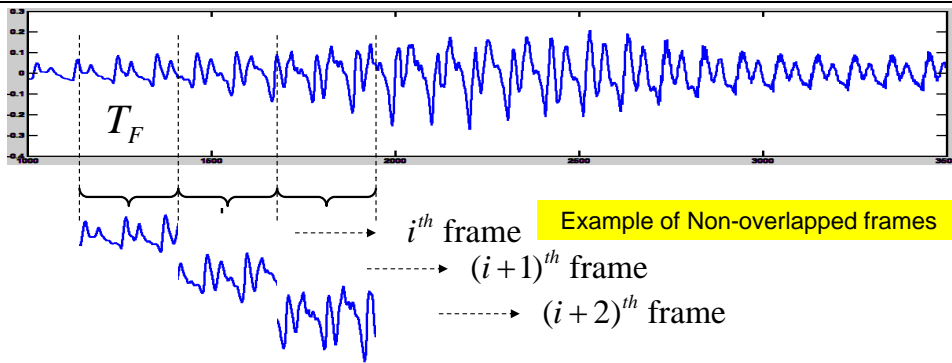
Observations

- When signal property stays fixed, the longer the section of data (larger L), the sharper the frequency resolution.
- The larger the number of points for the DFT, N, the denser the frequency axis sampling.
- However, the resolution (or separation) of sinusoids w.r.t. frequency is determined by the data length L, not by length of the N-pt DFT (with zero padding).
- When signal properties change with time, there are other considerations that limit the size of L; see case **B**.

Case B - The General Case

- DFT can also be used as a tool for finding the spectral composition of an arbitrary signal:
 - In general, we observe changing signal properties
 - Short-time** analysis framework
 - Analyze sequence one block (frame) of data at a time using DFT
 - Successive blocks of data may overlap
 - "SPECTROGRAM"**

FRAME = WINDOW of DATA



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A very long signal

- Four intervals
- Four spectrum lines

$$x[n] = x(n / f_s) = \begin{cases} 5 \cos(0.211\pi n) & 0 \leq n \leq 499 \\ 2 \cos(0.111\pi n) & 500 \leq n \leq 2999 \\ 2 \cos(0.8\pi n) & 3000 \leq n \leq 4999 \\ \frac{1}{2} \cos(0.4\pi n) & 5000 \leq n \leq 9999 \end{cases}$$

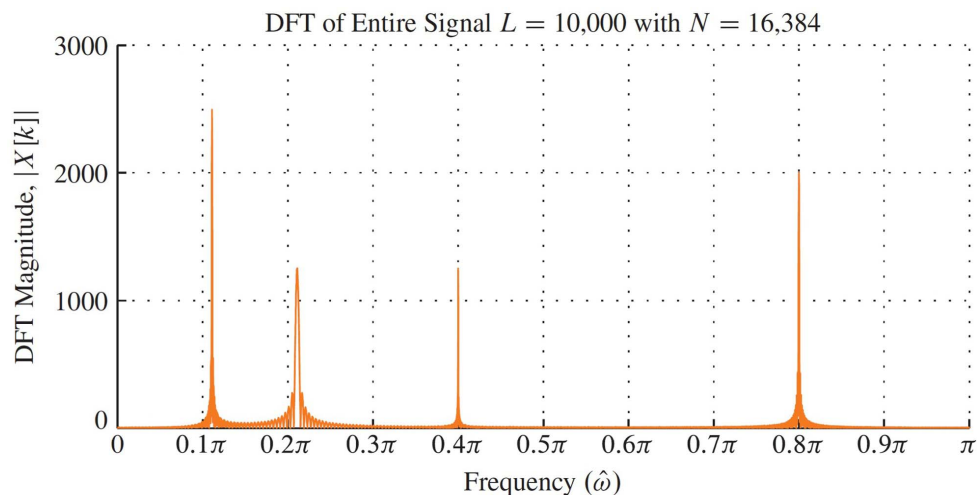
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DFT of a very Long Signal (Global)

- Length 10,000 signal, 16384-pt DFT



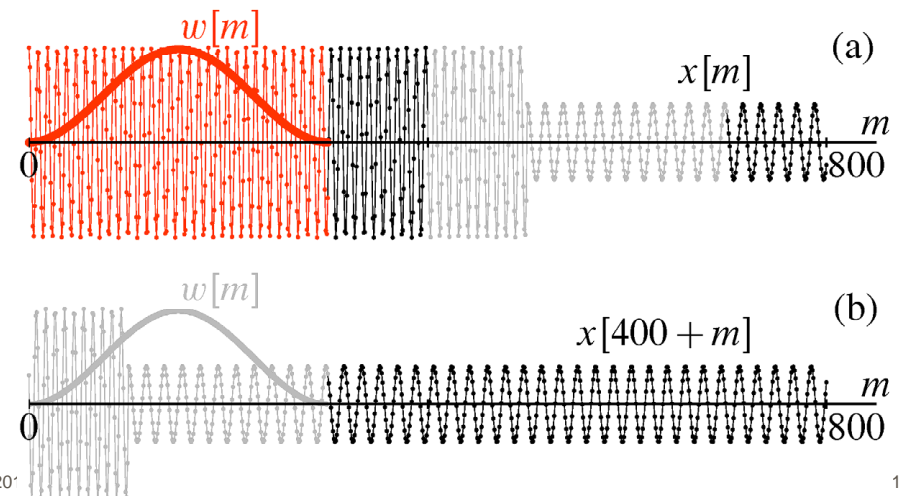
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Windowing a Long Signal

- Long signal is time-shifted into the domain of the window, $[0,300]$ in this case

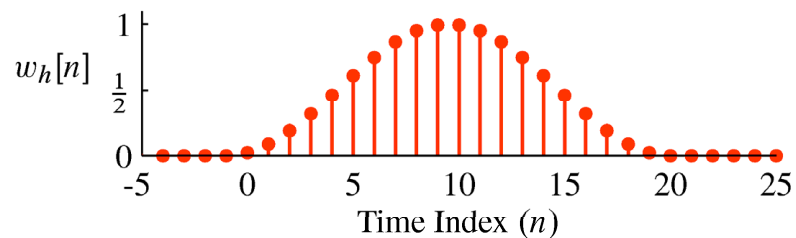


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Von Hann Window in Time Domain

- Plot of Length-20 von Hann window



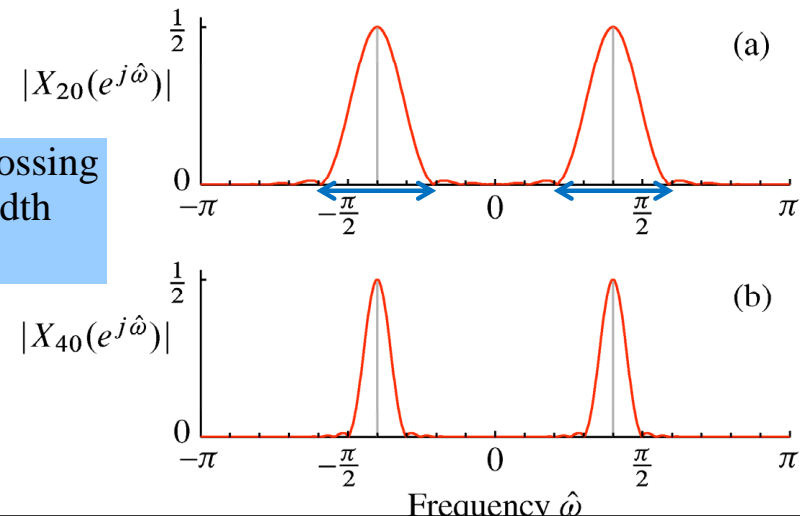
von Hann Window (Length L)

$$w_h[n] = \begin{cases} 0 & n < 0 \\ \frac{1}{2} - \frac{1}{2} \cos(2\pi(n+1)/(L+1)) & 0 \leq n < L \\ 0 & n \geq L \end{cases}$$

DTFT of Hann Window: Change Length

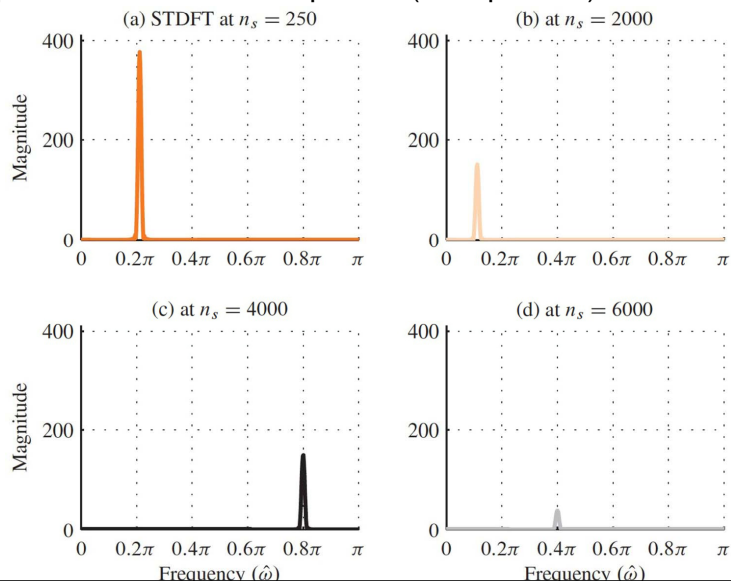
- DTFT (magnitude) of windowed sinusoid.
 - Length-20 Hann window vs. Length-40 Hann window

Zero - crossing
Peak Width
 $\Delta\hat{\omega} = \frac{8\pi}{L}$



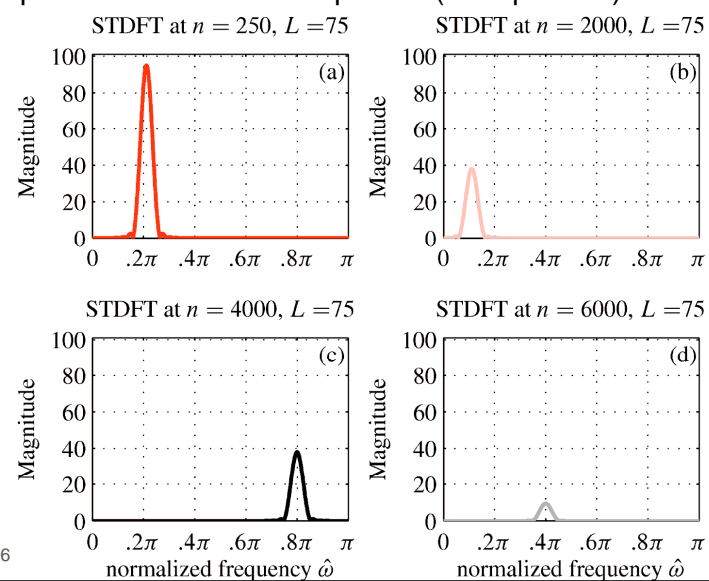
Short DFTs of Windowed Sections (L=301)

- 301-pt Hann window. 1024-pt DFT (zero-padded)



Short DFTs of Windowed Sections (L=75)

- 75-pt Hann window. 1024-pt DFT (zero-padded)



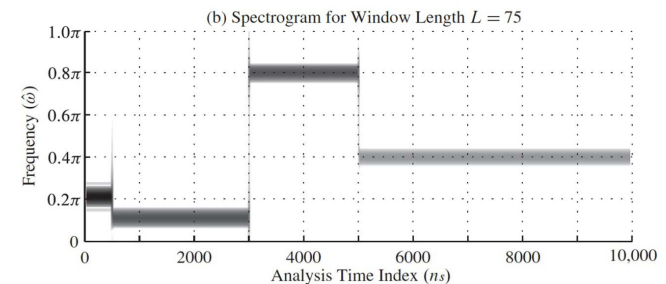
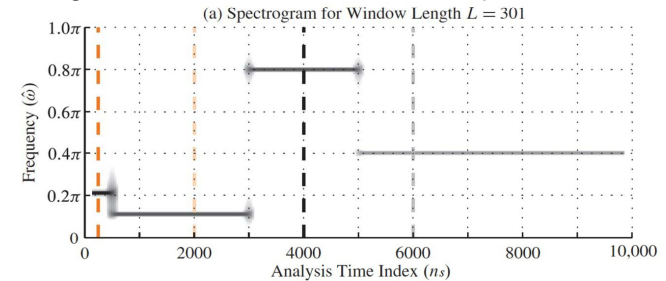
A very long signal

- Four intervals
- Four spectrum lines

$$x[n] = x(n/f_s) = \begin{cases} 5 \cos(0.211\pi n) & 0 \leq n \leq 499 \\ 2 \cos(0.111\pi n) & 500 \leq n \leq 2999 \\ 2 \cos(0.8\pi n) & 3000 \leq n \leq 4999 \\ \frac{1}{2} \cos(0.4\pi n) & 5000 \leq n \leq 9999 \end{cases}$$

Two Spectrograms of very Long Signal

- Window Lengths of L=301 and L=75; overlap 90%



MATLAB Spectrogram

- In MATLAB, use `S = spectrogram(X, WINDOW, NOVERLAP, NFFT, Fs, 'yaxis')`
- f_s F_s is the sampling frequency specified in samples/s
- WINDOW:
 - L - if a vector, extract segments of x of equal length, and then multiply window vector and data vector, element by element;
 - If an integer, same as above but use a Hamming window of the specified length
 - if unspecified, use default
- N_x NOVERLAP: ($<WINDOW$)
 - Number of overlapping data between successive frames
- N NFFT: Length of FFT

Resolution Test

- Closely spaced frequency components
 - How close can the lines be?
 - Depends on window (section) length
 - Inverse relationship

$$x[n] = x(n/f_s) = \begin{cases} \cos(0.2\pi n) + 1.5 \cos(0.227\pi n) & 0 \leq n \leq 4999 \\ \cos(0.2\pi n) & 5000 \leq n \leq 6999 \\ 3 \cos(0.6\pi n) + \cos(0.7\pi n) & 7000 \leq n \leq 9999 \end{cases}$$

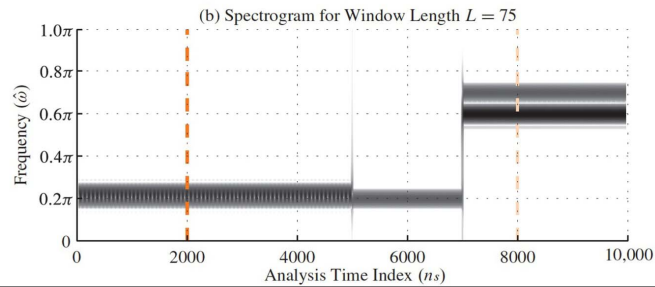
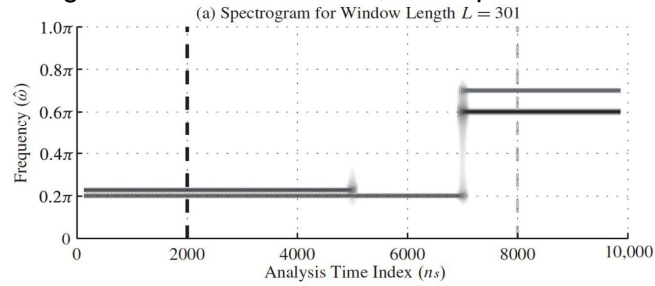
$$\cos(0.2\pi n) + 1.5 \cos(0.227\pi n) \Rightarrow \Delta \hat{\omega} = 0.027\pi = 8\pi / ?$$

$$3 \cos(0.6\pi n) + \cos(0.7\pi n) \Rightarrow \Delta \hat{\omega} = 0.1\pi = 8\pi / 80$$

$$\left(\frac{\Delta \hat{\omega}}{8\pi}\right) = \frac{1}{296}$$

Two Spectrograms: Resolution Test

- Window Lengths of $L=301$ and $L=75$, overlap 90%



Short DFTs of Windowed Sections

- 75-pt and 301-pt Hann windows. 1024-pt DFT (zero-padded)

