

DSP First, 2/e

Lecture 23 Frequency Response, $H(z)$, Poles and Zeros for IIR and FIR Systems

READING ASSIGNMENTS

- This Lecture:
 - Chapter 9, Sects. 9-5 and 9-6
 - Chapter 10, Sects. 10-5 and 10-7

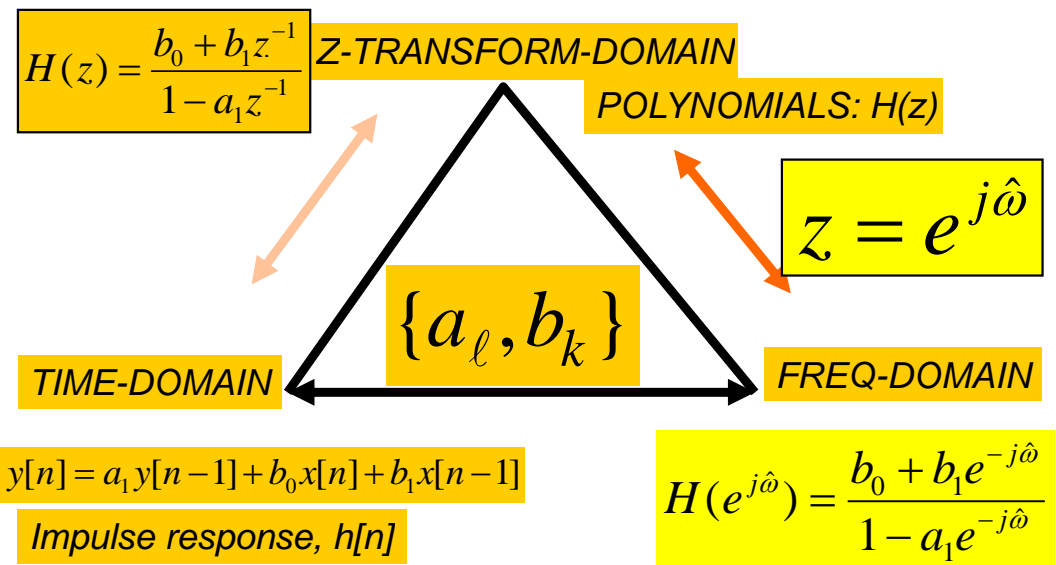
LECTURE OBJECTIVES

- ZEROS and POLES
- Relate $H(z)$ to FREQUENCY RESPONSE

$$H(e^{j\hat{\omega}}) = H(z) \Big|_{z=e^{j\hat{\omega}}}$$

- Four demos: PeZ, 3-Domain movies
 - Placing Poles and Zeros
- Bandpass Filters: IIR
- Nulling Filters: FIR Notch Filters: IIR

THREE DOMAINS: $H(e^{j\hat{\omega}})$



Motivation: Filter Design

- Some tasks/analysis easier in one domain
 - Freq domain: system response to sinusoids
 - Time domain: calculate output to any signal
 - Z-domain: given specs, build a filter
- Can we design a filter that removes DC and sinusoids at frequency $\hat{\omega} = \pi/3$?
- Z-domain reduces this to polynomial roots

H(z) = Rational Function

- First Order:

$$H(z) = \frac{b_0 + b_1 z^{-1}}{1 - a_1 z^{-1}}$$

- We can also study Second-Order Systems:

$$H(z) = \frac{b_0 + b_1 z^{-1} + b_2 z^{-2}}{1 - a_1 z^{-1} - a_2 z^{-2}} = \frac{B(z)}{A(z)}$$

- Numerator & Denominator Polynomials

POLES & ZEROS of H(z)

- Zeros of H(z), i.e., where is H(z)=0?
 - Look for Roots of Numerator Polynomial
- Poles of H(z), i.e., where is H(z)=infinity?
 - Look for Roots of Denominator Polynomial

$$H(z) = \frac{B(z)}{A(z)}, \text{ so } B(z_0) = 0 \Rightarrow H(z_0) = 0$$

if $A(z_0) \neq 0$

$$H(z) = \frac{B(z)}{A(z)}, \text{ so } A(z_0) = 0 \Rightarrow H(z_0) \rightarrow \infty$$

if $B(z_0) \neq 0$

Poles/Zeros of 1st-order H(z)

- Roots of Numerator & Denominator Polys:

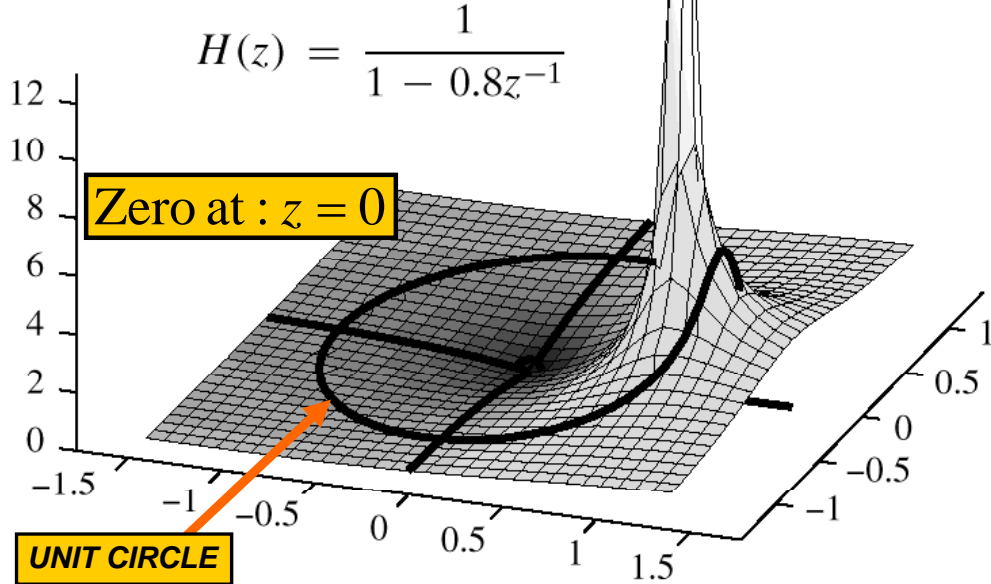
$$H(z) = \frac{1 + b_1 z^{-1}}{1 - 0.8 z^{-1}}$$

$$H(z) = \frac{z(1 + b_1 z^{-1})}{z(1 - 0.8 z^{-1})} = \frac{z + b_1}{z - 0.8}$$

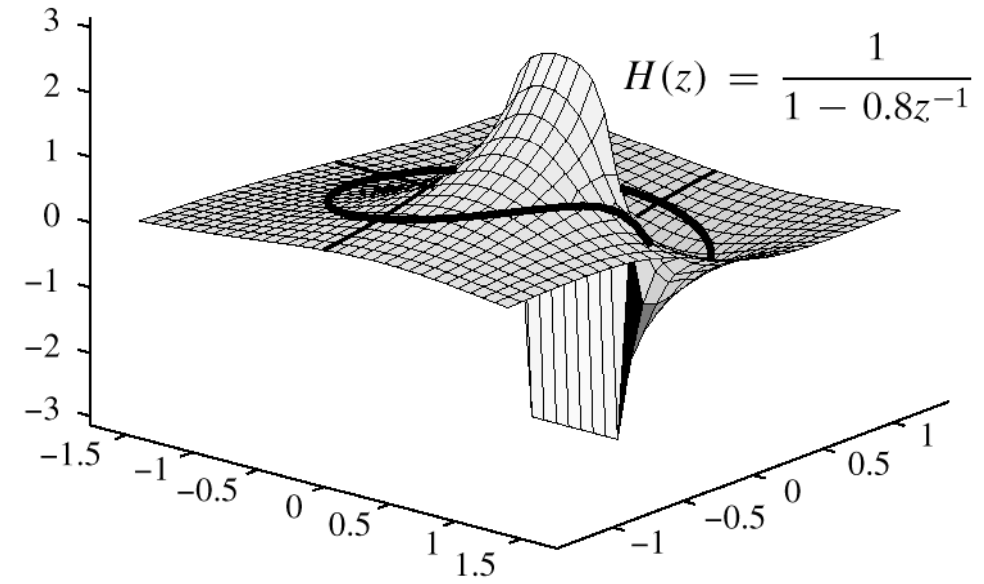
Pole at : $z = 0.8$

Zero at : $z = -b_1$

**3-D VIEWPOINT:
EVALUATE H(z) EVERYWHERE**



PHASE from 3-D PLOT



FREQ. RESPONSE from H(z)

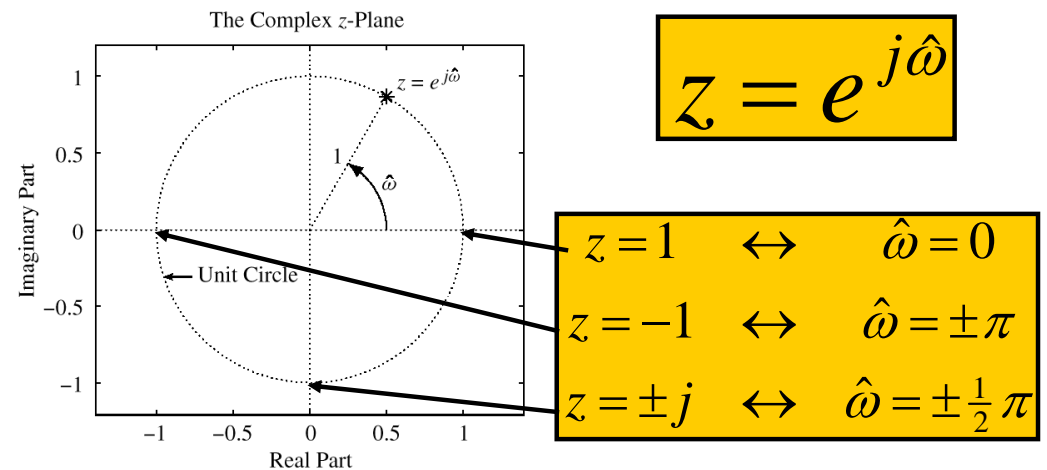
$H(e^{j\hat{\omega}}) = H(z) \Big|_{z=e^{j\hat{\omega}}}$

- Relate H(z) to FREQUENCY RESPONSE
- EVALUATE H(z) on the **UNIT CIRCLE**
 - ANGLE is same as FREQUENCY

$z = e^{j\hat{\omega}}$ (as $\hat{\omega}$ varies)
defines a **CIRCLE**, radius = 1

UNIT CIRCLE: RECAP

- MAPPING BETWEEN z and $\hat{\omega}$

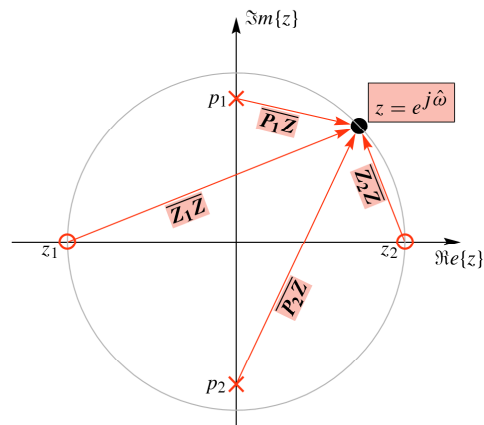


Frequency Response from poles and zeros

$$|H(e^{j\hat{\omega}})| = G \frac{|e^{j\hat{\omega}} - z_1| |e^{j\hat{\omega}} - z_2|}{|e^{j\hat{\omega}} - p_1| |e^{j\hat{\omega}} - p_2|}$$

$$|H(e^{j\hat{\omega}})| = G \frac{\overline{Z_1 Z} \cdot \overline{Z_2 Z}}{P_1 Z \cdot P_2 Z}$$

$$H(z) = G \frac{(z - z_1)(z - z_2)}{(z - p_1)(z - p_2)}$$



IIR H(z) example: two poles

- Poles just inside the unit circle (for stability)

$$H(z) = \frac{1}{1 + 0.97z^{-1} + 0.9409z^{-2}}$$

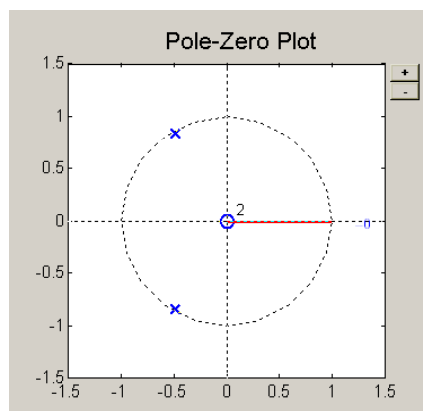
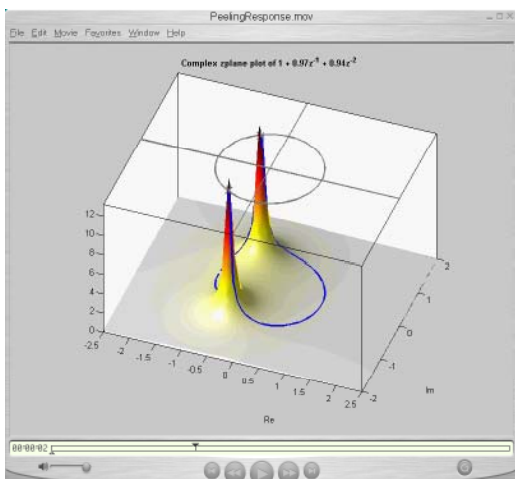
2 Poles : $z = 0.97e^{\pm j2\pi/3}$

2 Zeros : $z = 0,0$

- MATLAB: `roots()` and `poly()`
 - `roots([1, 0.97, 0.9409])`
 - `poly(0.97*exp(j*2*pi*[1,-1]/3))`

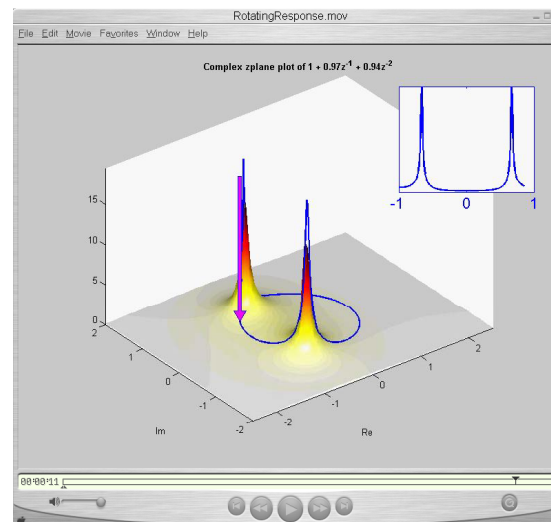
MOVIE for H(z) in 3-D

- POLES to H(z) to Frequency Reponse
- TWO POLES SHOWN

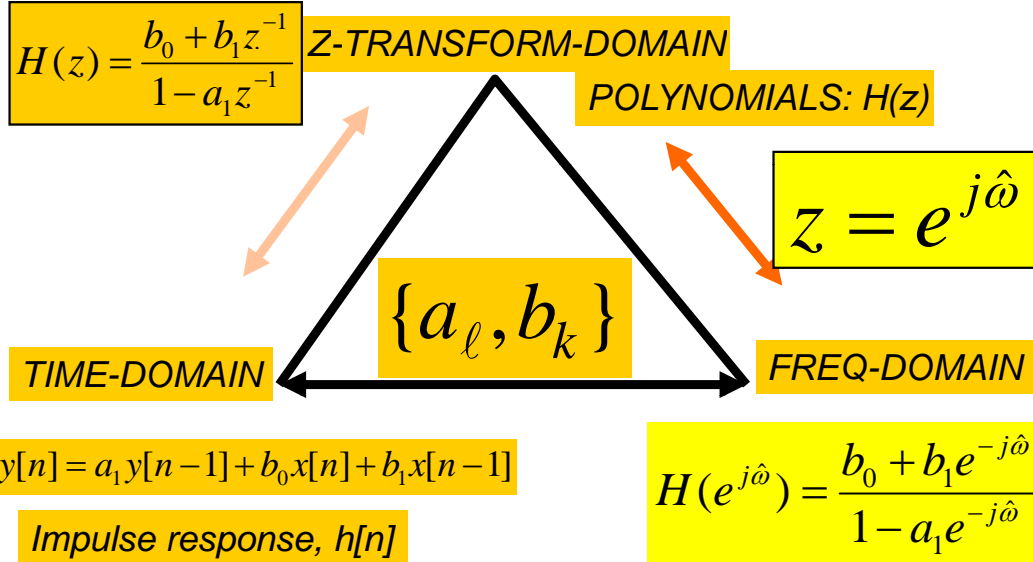


Frequency Response from H(z)

Walking around the Unit Circle



THREE DOMAINS: $H(e^{j\hat{\omega}})$

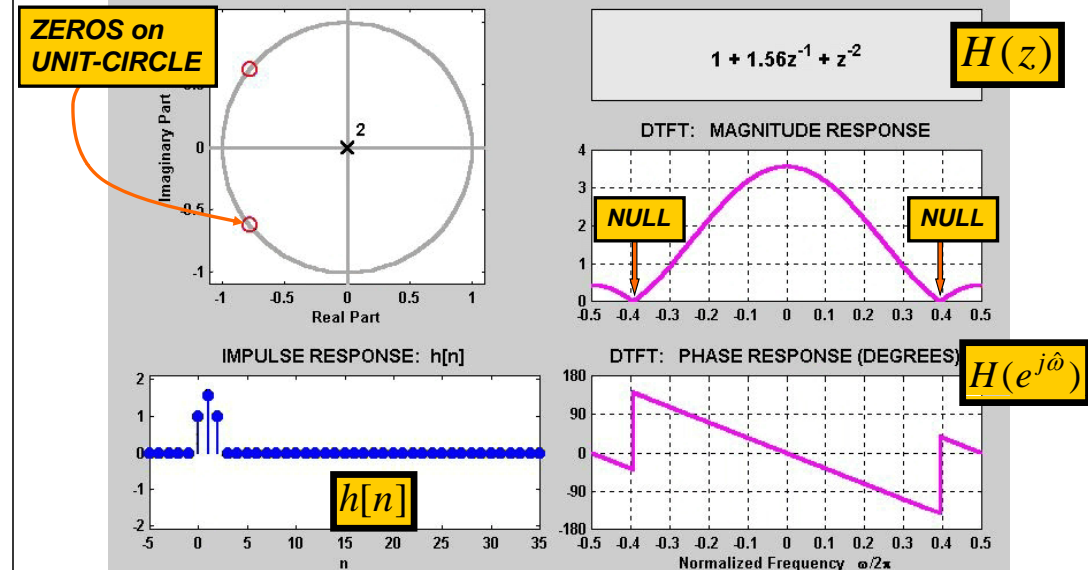


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3 DOMAINS MOVIE: FIR

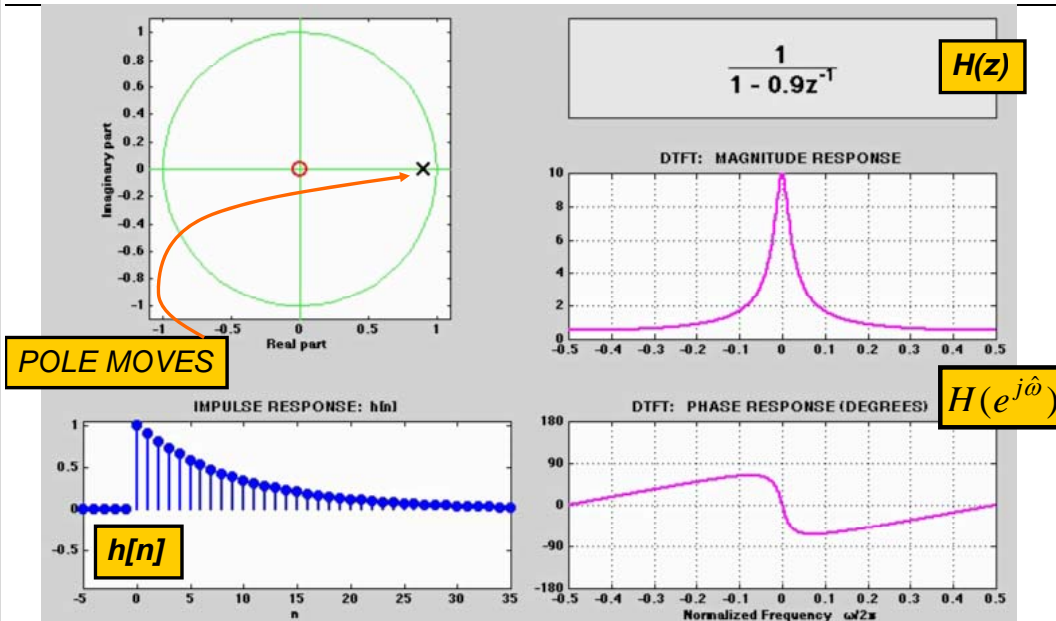


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3 DOMAINS MOVIE: IIR



7 IIR MOVIES @ WEBSITE

▪ http://dspfirst.gatech.edu/chapters/08feedback/demos/3_domain/index.html

3 DOMAINS MOVIES: IIR Filters

- One pole moving and a zero at the origin
- One pole and one zero; both moving
- Two complex-conjugate poles moving radially
- Two complex-conjugate poles moving in angle
- Movement of a zero in a two-pole Filter
- Radial Movement of Two out of Four Poles
- Angular Movement of Two out of Four Poles

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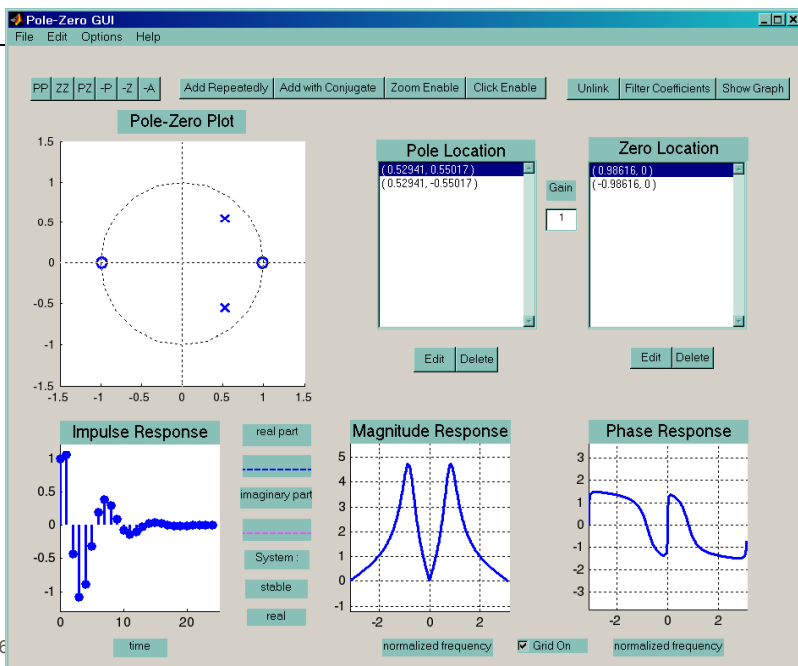
Reminder: 4 FIR MOVIES @ WEBSITE

- http://dspfirst.gatech.edu/chapters/08feedback/demos/3_domain/index.html
- 3 DOMAINS MOVIES: FIR Filters
 - Two zeros moving around UC and inside
 - Three zeros; one held fixed at $z = -1$
 - Ten zeros; 9 equally spaced around UC; one moving
 - Ten zeros; 8 equally spaced around UC; two moving

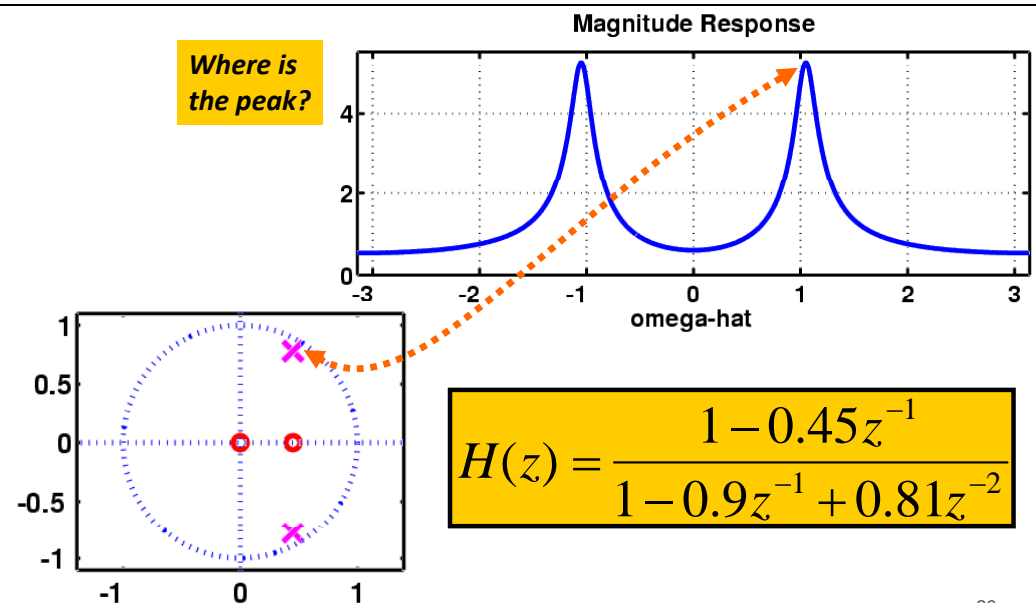
Remove Interference

- Design a NOTCH filter (Find a_k and b_k)
 - To **Reject** completely 0.7π
 - This is NULLING
 - Zeros on UC **2 Zeros: $z = e^{\pm j0.7\pi}$**
 - Make the frequency response magnitude FLAT away from the notch. **2 Poles: $z = 0.97e^{\pm j0.7\pi}$**
 - Use poles at the **same angle**
- Z-POLYNOMIALS provide the TOOLS
 - PEZDEMO GUI

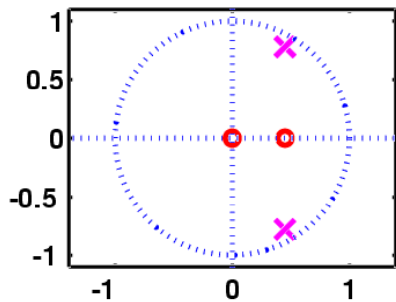
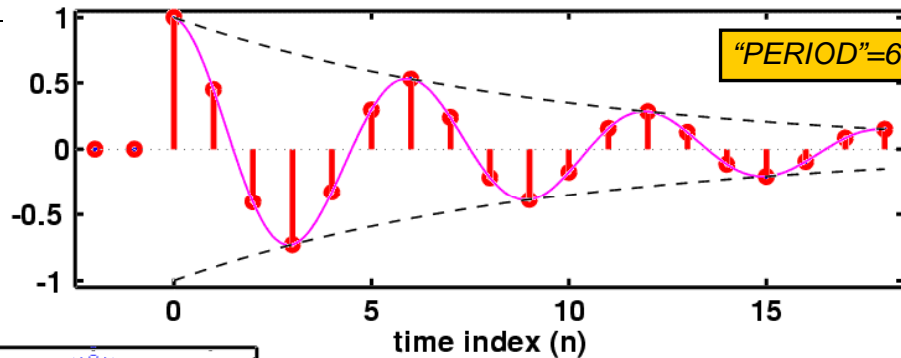
PeZ Demo: Pole-Zero Placing



Complex POLE-ZERO PLOT

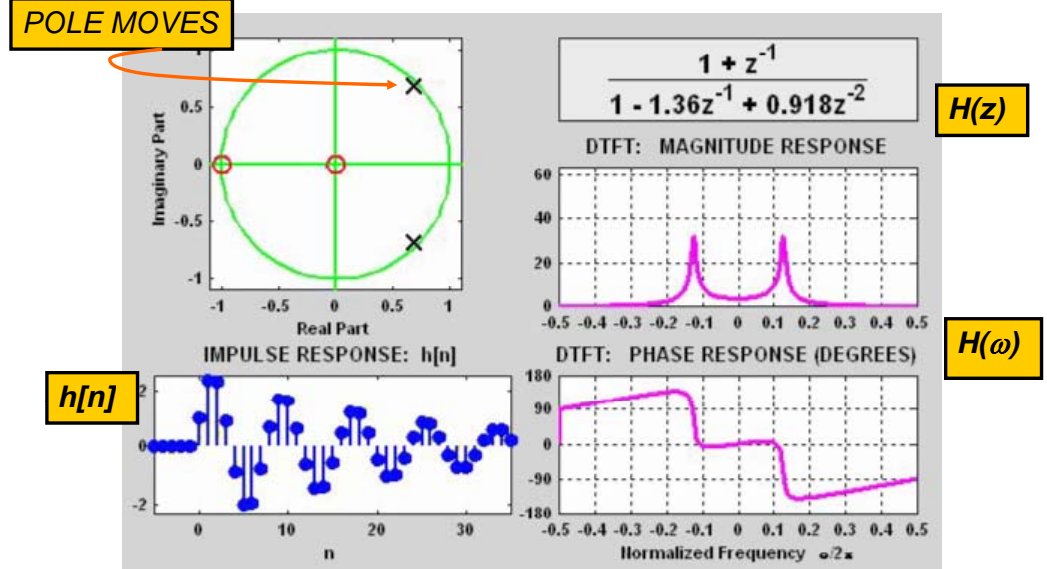


h[n]: Decays & Oscillates



$$h[n] = (0.9)^n \cos(\pi n / 3) u[n]$$

3 DOMAINS MOVIE: IIR



SINUSOIDAL RESPONSE

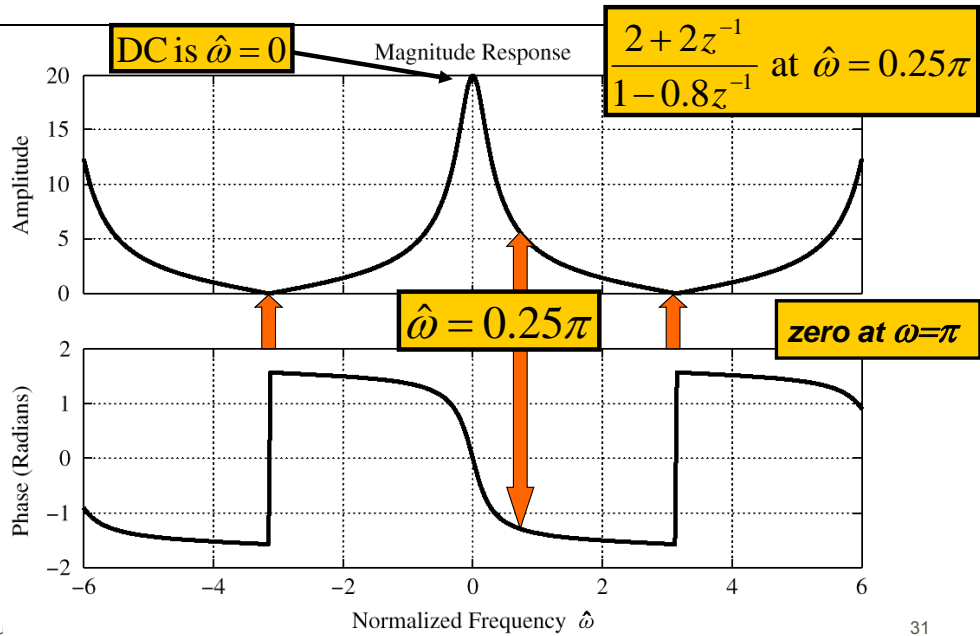
- $x[n] = \text{SINUSOID} \Rightarrow y[n]$ is SINUSOID
- Get MAGNITUDE & PHASE from $H(z)$

if $x[n] = e^{j\hat{\omega}n}$
 then $y[n] = H(e^{j\hat{\omega}}) e^{j\hat{\omega}n}$
 where $H(e^{j\hat{\omega}}) = H(z) \Big|_{z=e^{j\hat{\omega}}}$

POP QUIZ

- Given: $H(z) = \frac{2 + 2z^{-1}}{1 - 0.8z^{-1}}$
- Find the Impulse Response, $h[n]$
- Find the output, $y[n]$
 - When $x[n] = \cos(0.25\pi n)$

Evaluate FREQ. RESPONSE



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POP QUIZ: Eval Freq. Resp.

- Given: $H(z) = \frac{2 + 2z^{-1}}{1 - 0.8z^{-1}}$
- Find output, $y[n]$, when $x[n] = \cos(0.25\pi n)$
 - Evaluate at $z = e^{j0.25\pi}$

$$H(z) = \frac{2 + 2\left(\frac{\sqrt{2}}{2} - j\frac{\sqrt{2}}{2}\right)}{1 - 0.8e^{-j0.25\pi}} = 5.182e^{-j1.309}$$

$$y[n] = 5.182 \cos(0.25\pi n - 0.417\pi)$$

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